

INVESTICE DO ROZVOJE VZDĚLÁVÁNÍ

Robustné PID-regulátory s obmedzeniami

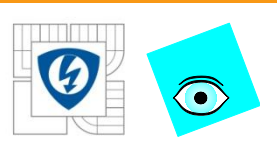
Robust Constrained PID Control

DC0: Robust I_0 and FI_0 controller design

prof. Ing. Mikuláš Huba, Ph.D.

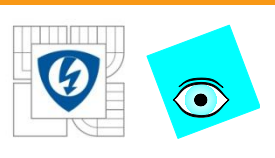
Ing. Peter Ťapák, Ph.D.

Tato prezentace je spolufinancována Evropským sociálním fondem a státním rozpočtem České republiky.



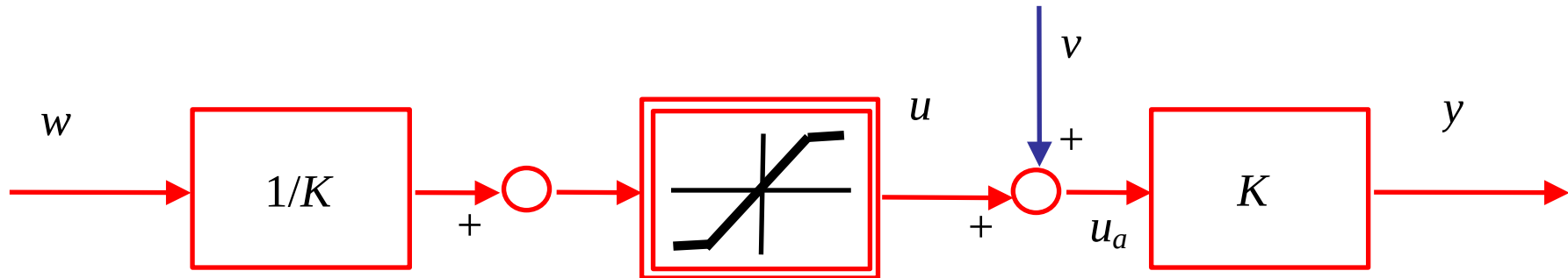
Contents : I_0 and FI_0 controllers

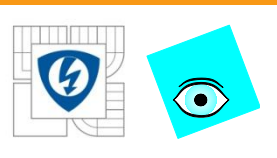
- Introduction
- Static feedforward control, FI_0 and I_0 controllers
- Generic and equivalent structures for compensation of input and output disturbances
- Nonmodelled dynamics
- Performance Portrait and Robustness Characteristics
- History



Input Disturbance Compensation

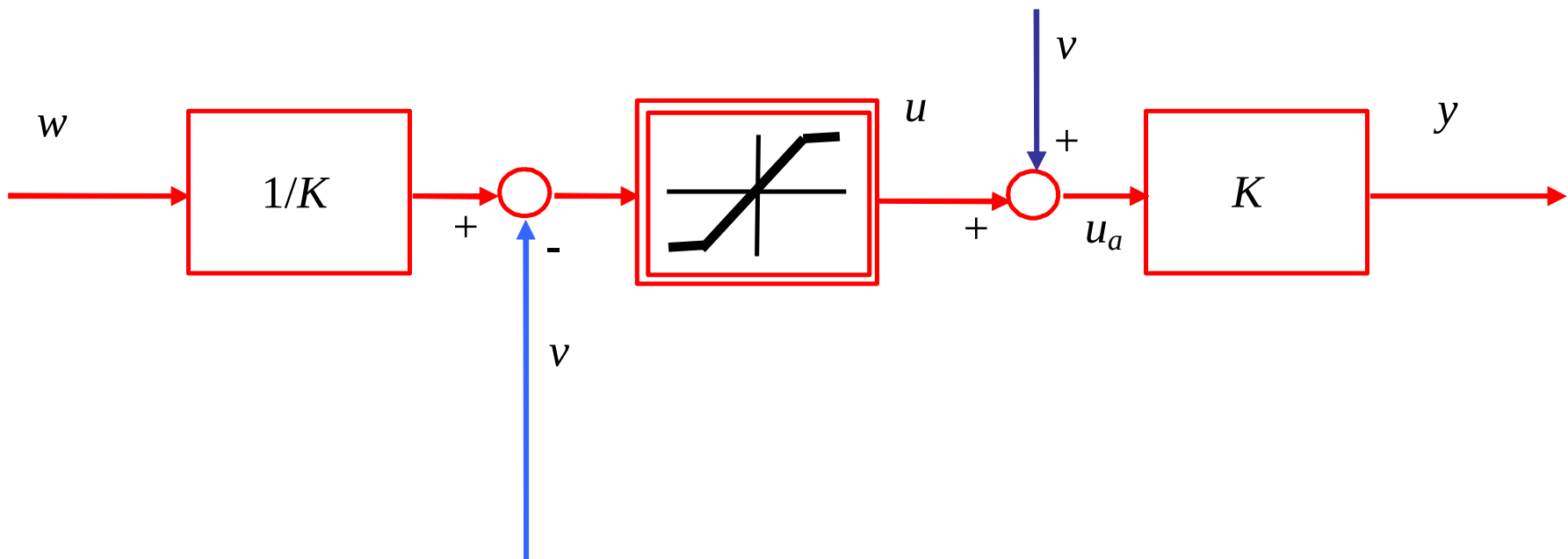
- Static feedforward control





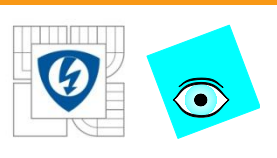
Input Disturbance (ID) Compensation

Static feedforward control with compensation of the input disturbance



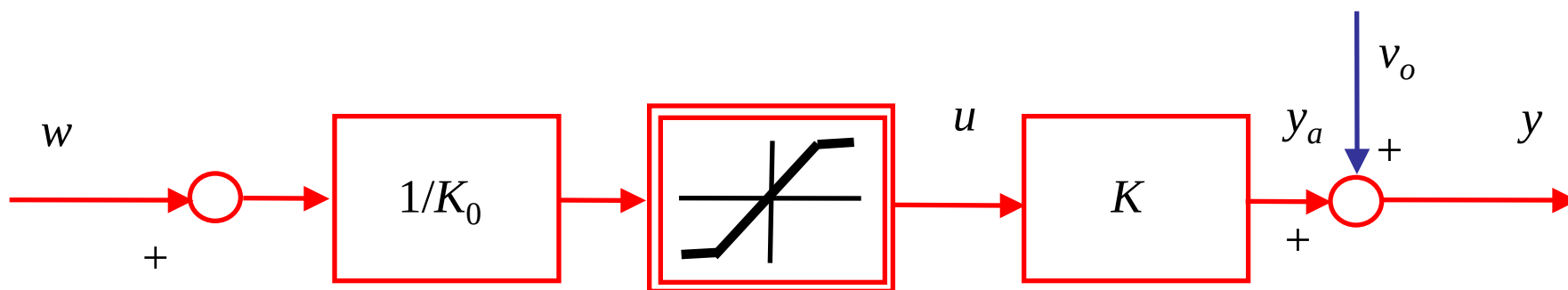
11.3.2011

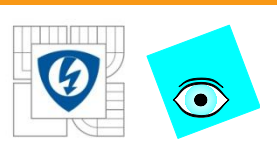
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Output Disturbance (OD) Compensation

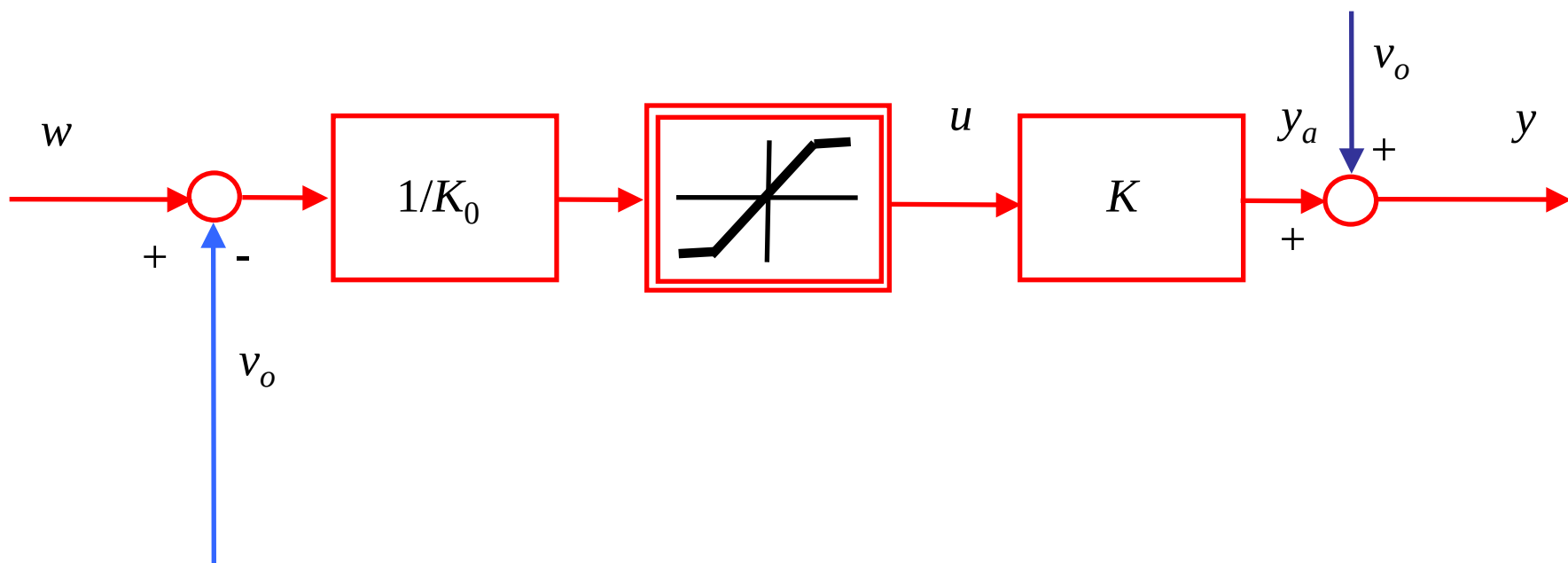
Static feedforward control





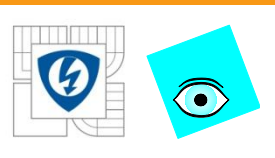
Output Disturbance Compensation

Static feedforward control with compensation of the output disturbance



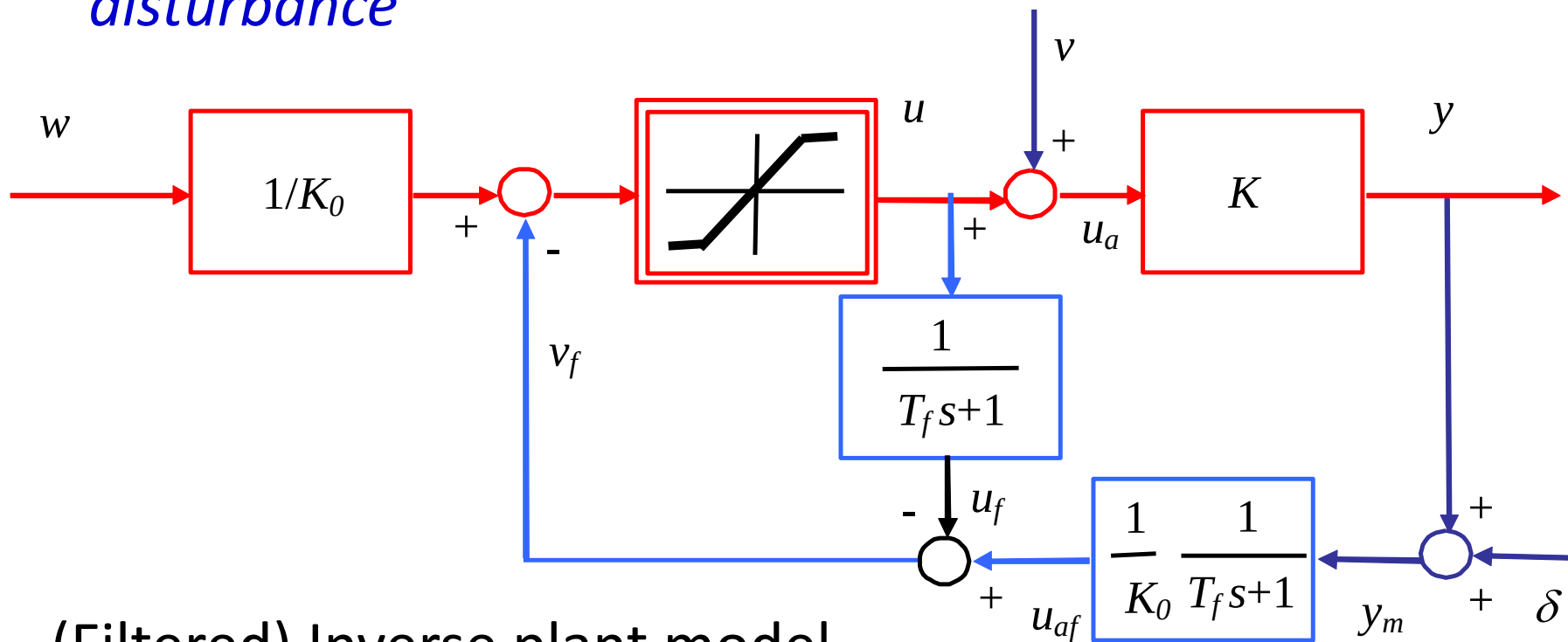
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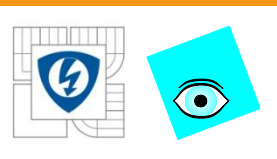


I_0 -ID controller – generic structure

Static feedforward control with compensation of input disturbance

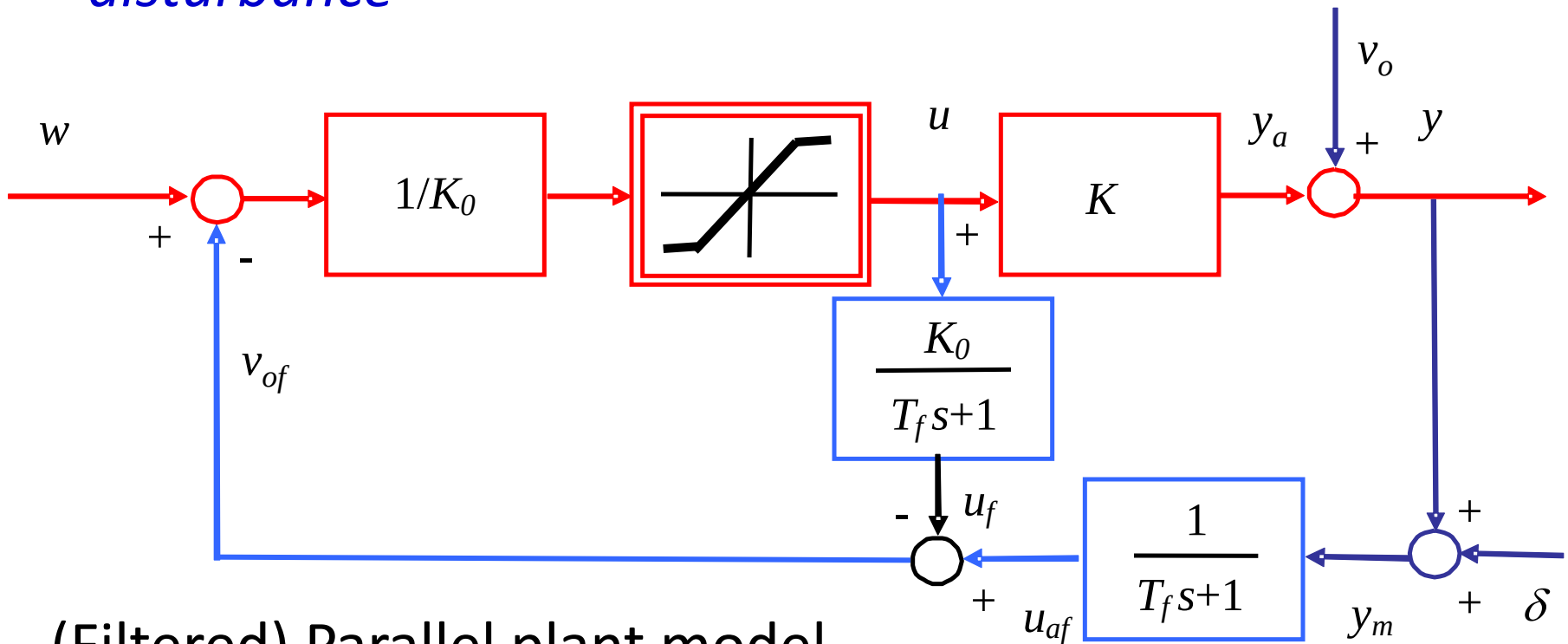


(Filtered) Inverse plant model

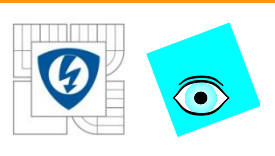


I_0 -OD controller – generic structure

Static feedforward control with compensation of output disturbance

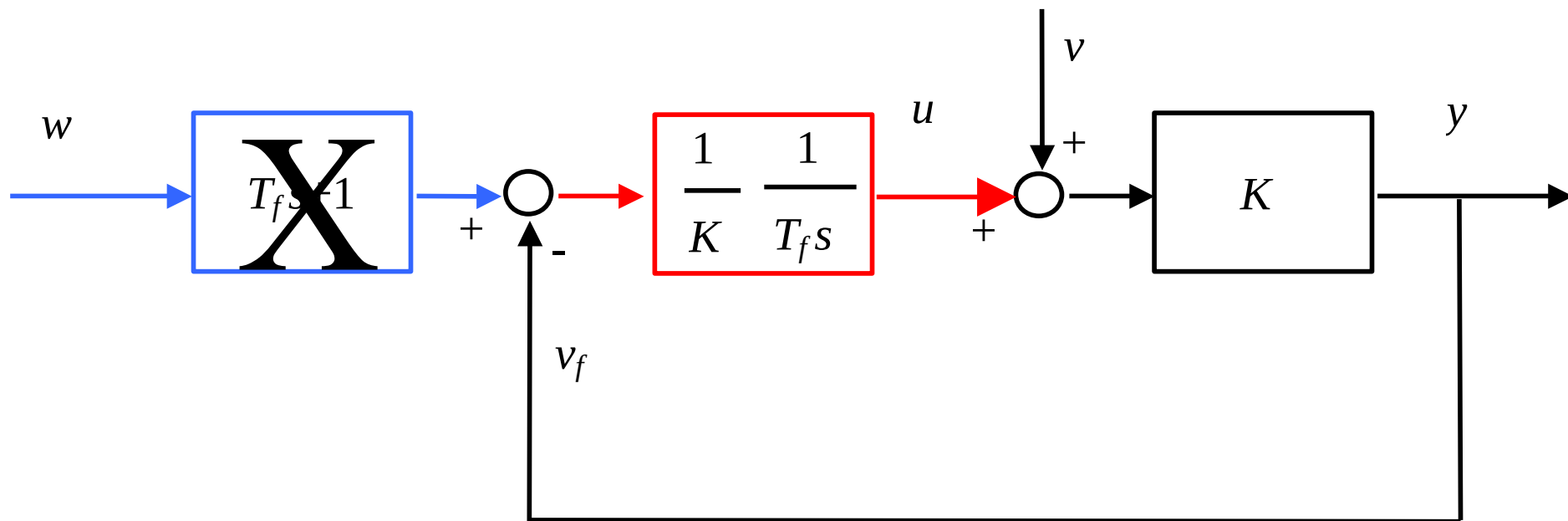


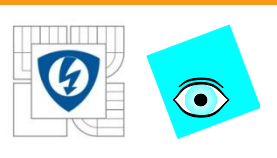
(Filtered) Parallel plant model



I_0 -ID controller - equivalent structure

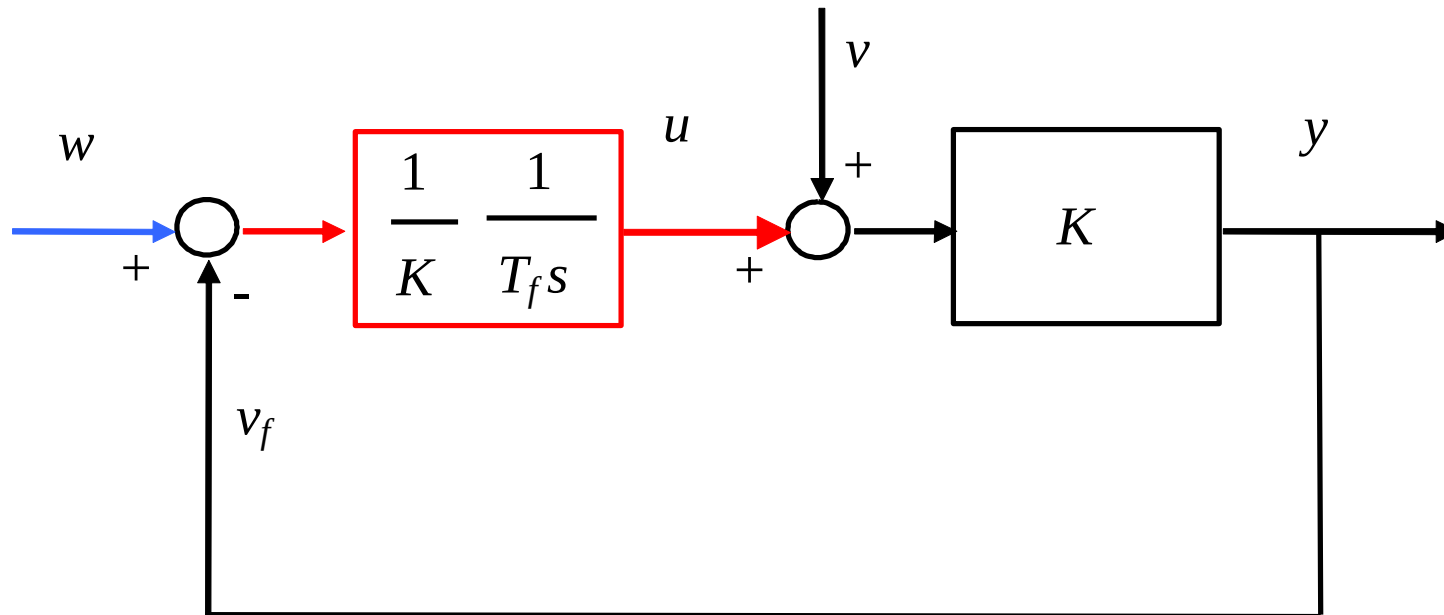
- explicit integrator
- Omitting prefilter = I-controller = FI_0 – controller

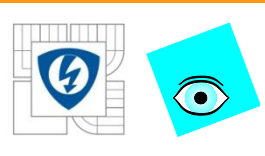




FI_0 controller = I controller

Filtered response continuous for $t = 0$





Explicit integrator

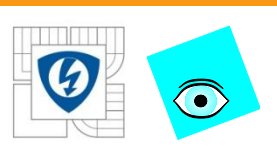
Involved in hydraulic & electromechanical actuators – mechanical constraints for integration



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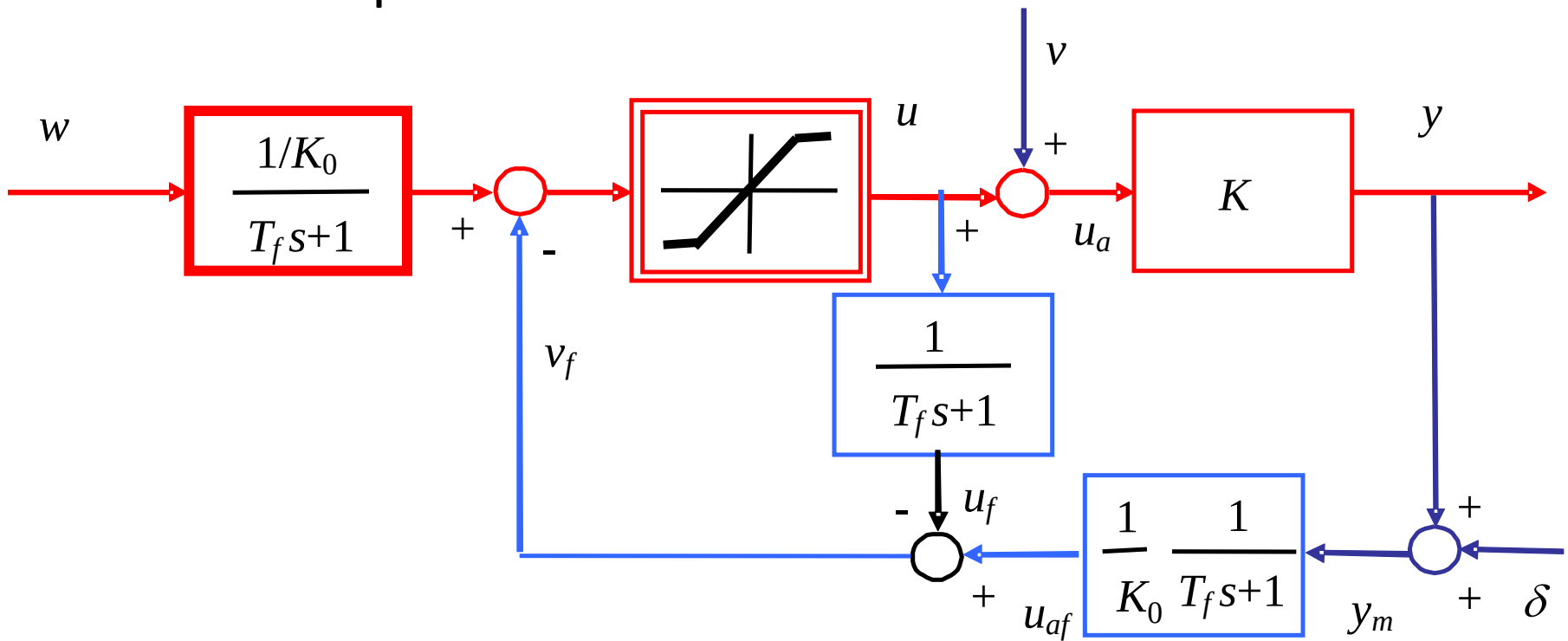
INVESTICE DO ROZVOJE VZDĚLÁVÁNÍ

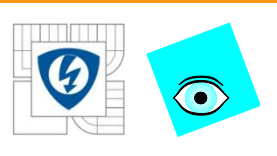




FI₀-ID-controller - generic fundamental structure

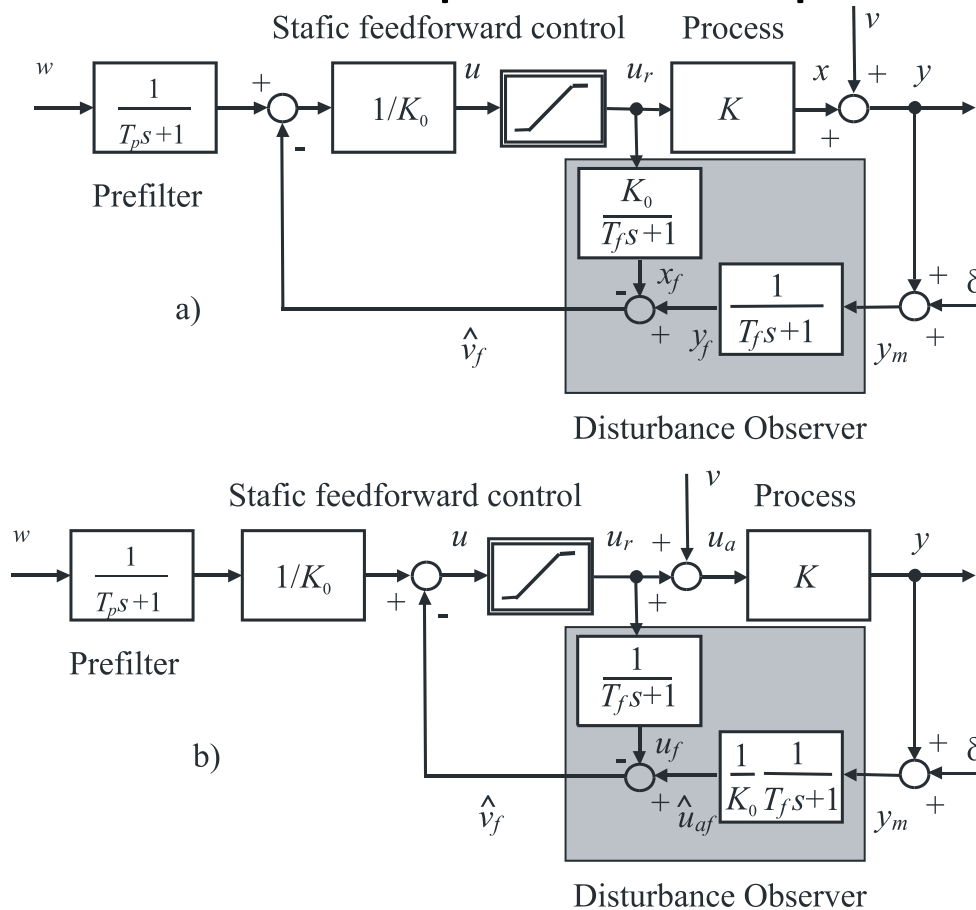
Omitting ideal prefilter in the equivalent scheme =
Introducing low-pass prefilter into generic scheme =
filtered response continuous for $t = 0$

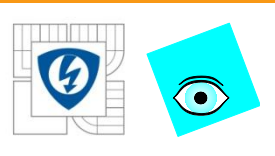




FI_0 -controllers (generic fundamental structures)

Two basic modifications for input and output disturbances





Influence of Nonmodelled Dynamics

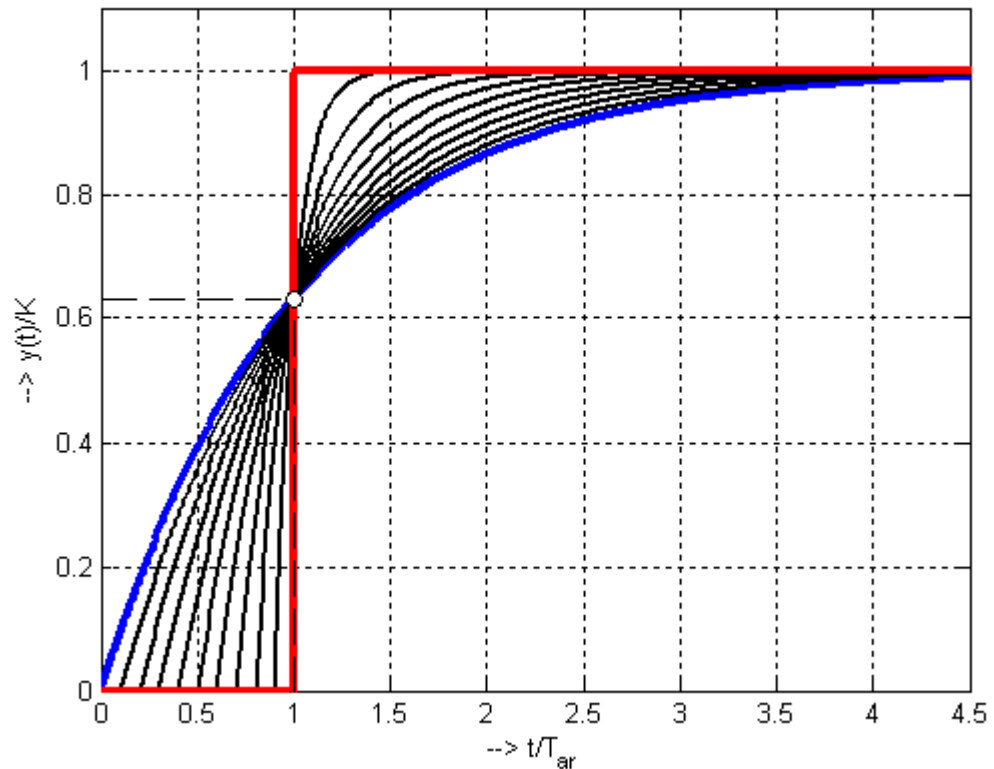
Approximation for the non-modeled dynamics
(not considered in deriving controller equation/structure)

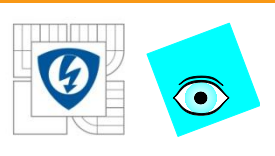
$$F_{nd}(s) = \frac{e^{-T_d s}}{1 + T_a s} ; T_{ar} = T_d + T_a$$

Several methods for
identification of T_{ar}
available

T_{ar} = time, at which the
normed step response
reaches 63% of the
steady state value

Different Step Responses of the FOPDT system with $T_{ar} = \text{const}$





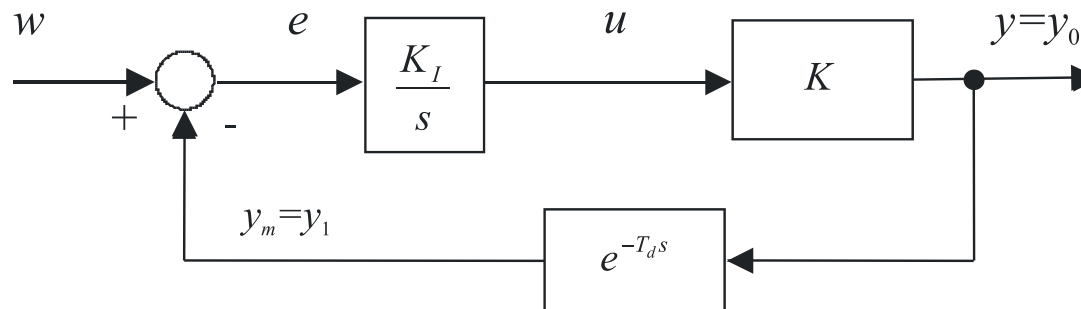
FI₀ controller+dead time T_d

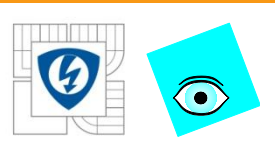
The dominant dynamics (memoryless plant) determines the controller structure

Usability limits and control quality of this structure depend on the nonmodelled dynamics

Dead time T_d – frequently used approximation of the nonmodelled dynamics

Dead time influence for the plant outputs y_0 and y_1 may be evaluated analytically or using the performance portrait method





FI₀ controller+dead time T_d

Closed loop transfer functions

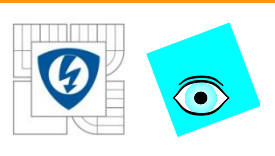
$$F_{w0}(s) = \frac{Y_0(s)}{W(s)} = \frac{K_I K e^{T_d s}}{s e^{T_d s} + K_I K} ; K_I = \frac{1}{K_0 T_f}$$
$$F_{w1}(s) = \frac{Y_1(s)}{W(s)} = \frac{K_I K}{s e^{T_d s} + K_I K}$$

Normalized variables

$$p = T_d s ; \kappa = \frac{K_0}{K} ; \Omega = \frac{T_d}{T_f} ;$$

Normalized transfer functions

$$F_{w0}(p) = \frac{e^p \Omega / \kappa}{p e^p + \Omega / \kappa} ; F_{w1}(p) = \frac{\Omega / \kappa}{p e^p + \Omega / \kappa}$$



FI₀ controller+dead time T_d

Charakteristic polynomial

$$A(p) = pe^p + \Omega / \kappa$$

Conditions of the double real pole p_0

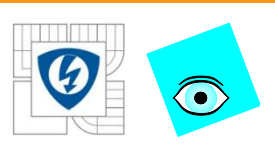
$$A(p_0) = 0 ; \dot{A}(p_0) = 0$$

Solution

$$\begin{aligned} \dot{A}(p) &= (p+1)e^p \Rightarrow p_0 = -1 \\ q &= \Omega / \kappa = \exp(-1); \tau = 1 / q = \exp(e) = 2.71... \end{aligned}$$

Optimal (analytical) tuning

$$\Omega = \kappa \exp(-1); \kappa = K_0 / K; \Omega = T_d / T_f$$



FI₀ controller+dead time T_d

Characteristic polynomial

$$A(p) = pe^p + \Omega / \kappa$$

Stability border $A(j\omega) = 0$

Stability border conditions

$$A(j\omega) = j\omega e^{j\omega} + \Omega / \kappa$$

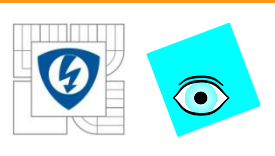
$$j\omega \cos \omega - \omega \sin \omega + \Omega / \kappa = 0 \Rightarrow \omega = \pm \pi / 2$$

Critical tuning

$$q = \Omega / \kappa = \pi / 2$$

Critical pole

$$\alpha_{crit} = -1/T_{f,crit} = -(K_0 / K)\pi / (2T_d)$$



FI_0 controller+dead time T_d

Performance Portrait
tolerated deviations
of MO & NO ε :

0 (red)

0.00001

0.0001

0.001

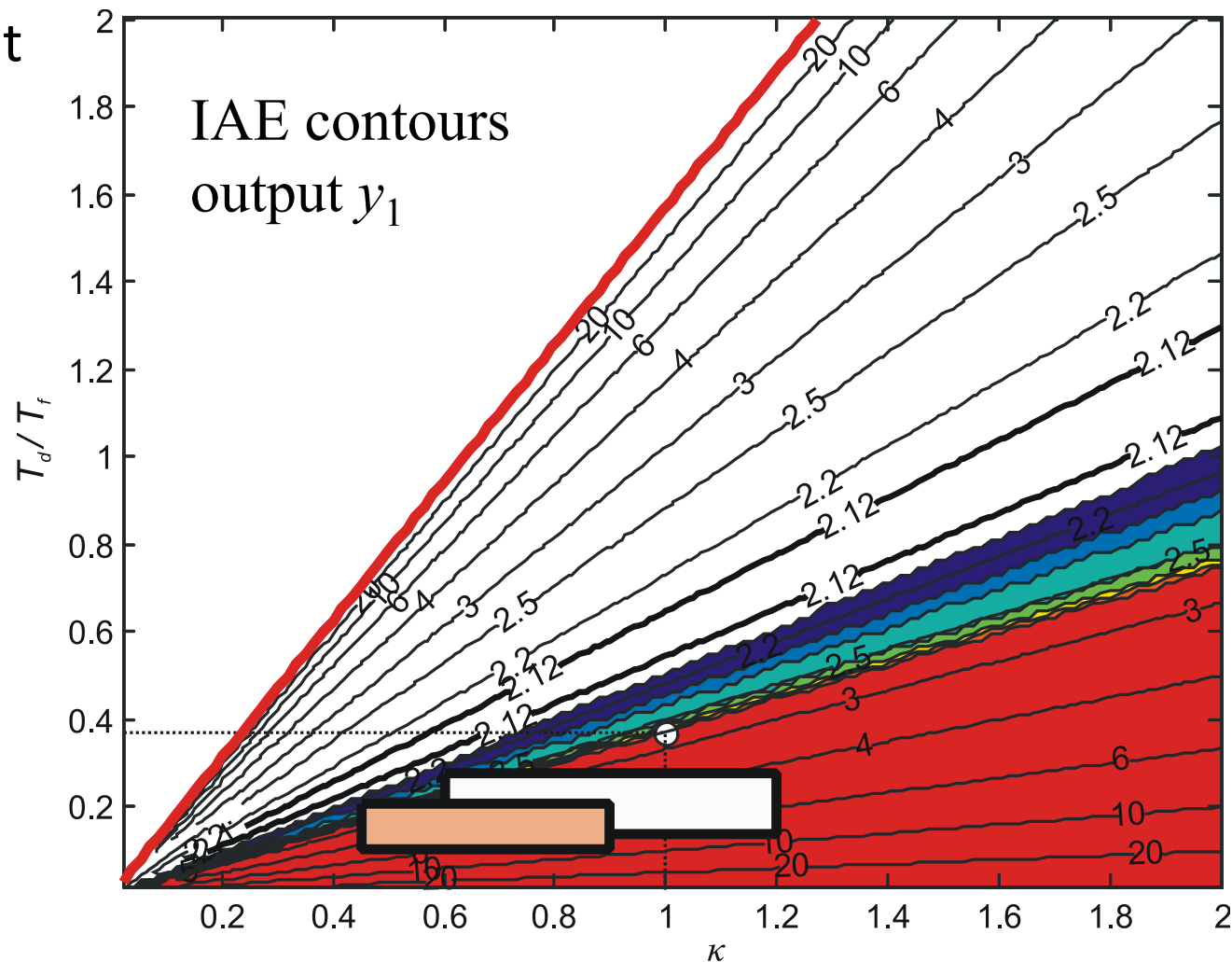
0.01

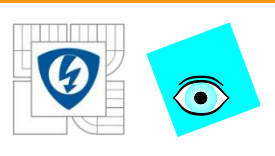
0.02

0.05 (darkblue)+

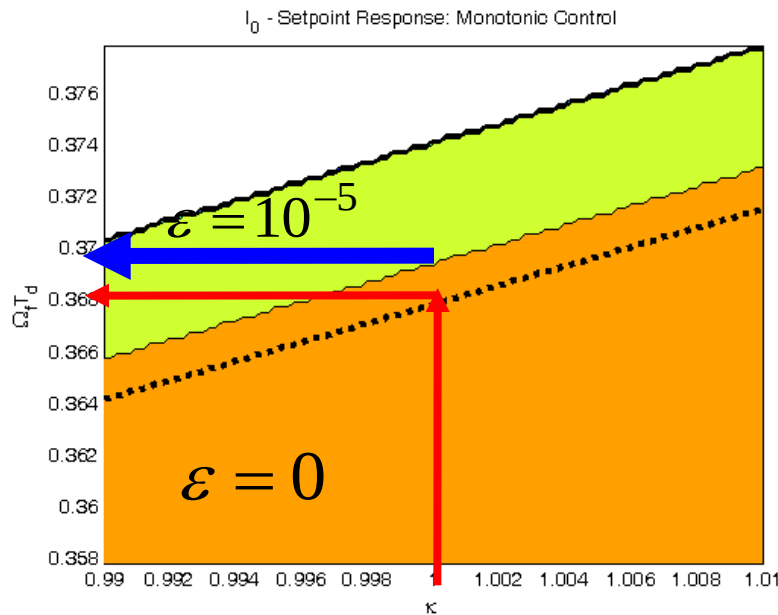
uncertainty boxes
with respect to

κ and T_d/T_f





FI_0 controller+dead time T_d

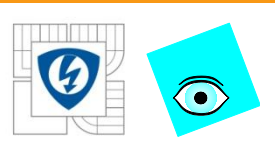


Weakened and strict MO
Experimentally determined MO
border different from the
aperiodicity border

$$\Omega = \kappa \exp(-1); \text{dotted}$$

$$\kappa = K_0 / K; \Omega = T_d / T_f$$

$100\varepsilon_y$ [%]	10	5	4.04	2	1	0.1	0.01	0.001	0	0
$\tau = \kappa / \Omega$	1.724	1.951	2.0	2.162	2.268	2.481	2.571	2.625	2.703	2.718
$q = \Omega / \kappa$	0.580	0.515	0.5	0.465	0.441	0.403	0.389	0.381	0.37	0.368
IAE_0 / T_d	1.105	1.147	1.17	1.240	1.314	1.486	1.571	1.625	1.703	1.718
IAE_1 / T_d	2.105	2.147	2.17	2.240	2.314	2.486	2.571	2.625	2.703	2.718



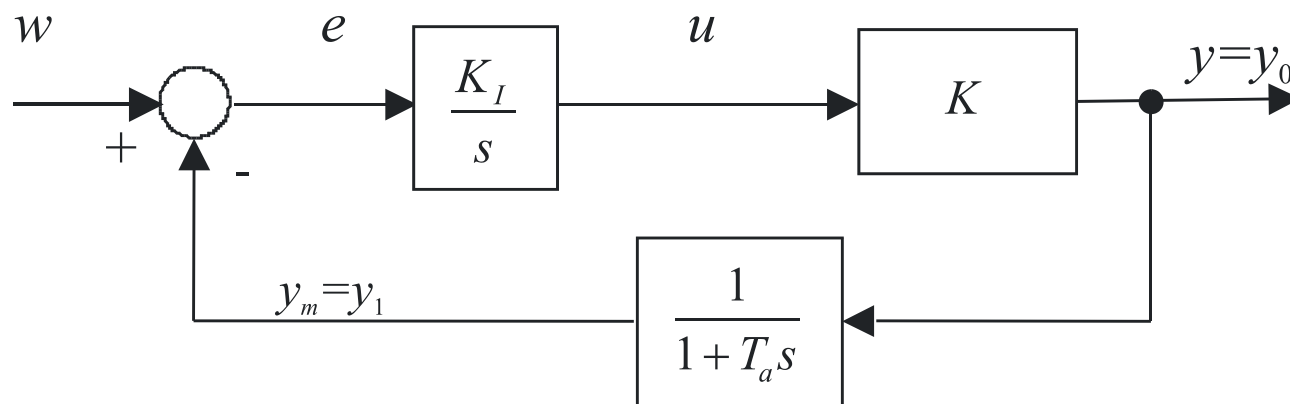
FI_0 controller+time constant T_a

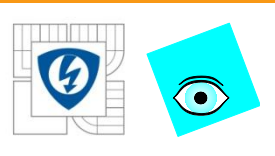
Closed loop transfer functions

$$F_{w0}(s) = \frac{Y_0(s)}{W(s)} = \frac{(1+T_a s)K_I K}{s(1+T_a s) + K_I K} = \frac{(1+T_a s)\Omega_f / \kappa}{s(1+T_a s) + \Omega_f / \kappa} = \frac{B_0(s)}{A(s)};$$

$$F_{w1}(s) = \frac{Y_1(s)}{W(s)} = \frac{K_I K}{s(1+T_a s) + K_I K} = \frac{\Omega_f / \kappa}{s(1+T_a s) + \Omega_f / \kappa} = \frac{B_1(s)}{A(s)};$$

$$\Omega_f = 1/T_f; \quad \kappa = K_0 / K$$





FI₀ controller+time constant T_a

Closed loop transfer functions

$$F_{w0}(p) = \frac{(1+p)\Omega / \kappa}{p(1+p) + \Omega / \kappa} = \frac{B_0(p)}{A(p)} ;$$

$$F_{w1}(p) = \frac{\Omega_f / \kappa}{p(1+p) + \Omega / \kappa} = \frac{B_1(p)}{A(p)} ;$$

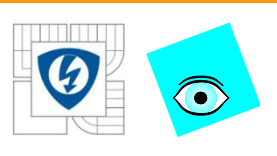
$$\Omega = T_a / T_f ; \quad \kappa = K_0 / K$$

Double real closed loop pole

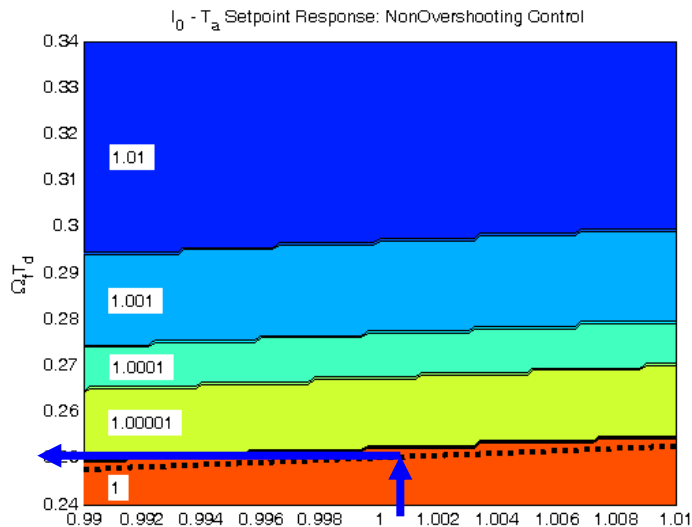
$$\tau = \kappa / \Omega = 4$$

Critical tuning

$$\Omega / \kappa \rightarrow 0$$



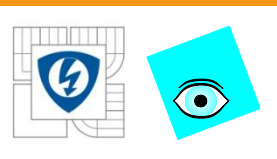
FI_0 controller+time constant T_a



Weakened and strict monotonicity
Experimentally determined
monotonicity border
different from the aperiodicity border
based on the DRDP

$$\kappa = 4\Omega$$

$100\varepsilon_y$ [%]	κ 10	5	2	1	0.1	0.01	0.001	0	0
$y_0 : \tau = \kappa / \Omega$	1.754	2.169	2.604	2.857	3.367	3.597	3.731	3.968	4
$q = \Omega / \kappa$	0.570	0.461	0.384	0.350	0.297	0.278	0.268	0.252	0.25
IAE_0 / T_a	1.348	1.510	1.760	1.941	2.376	2.598	2.732	2.968	3
$y_1 : \tau = \kappa / \Omega$	1.398	1.908	2.433	2.724	3.311	3.571	3.717	3.968	4
$q = \Omega / \kappa$	0.715	0.524	0.411	0.367	0.302	0.280	0.269	0.252	0.25
IAE_1 / T_a	1.921	2.221	2.581	2.806	3.321	3.573	3.717	3.968	4

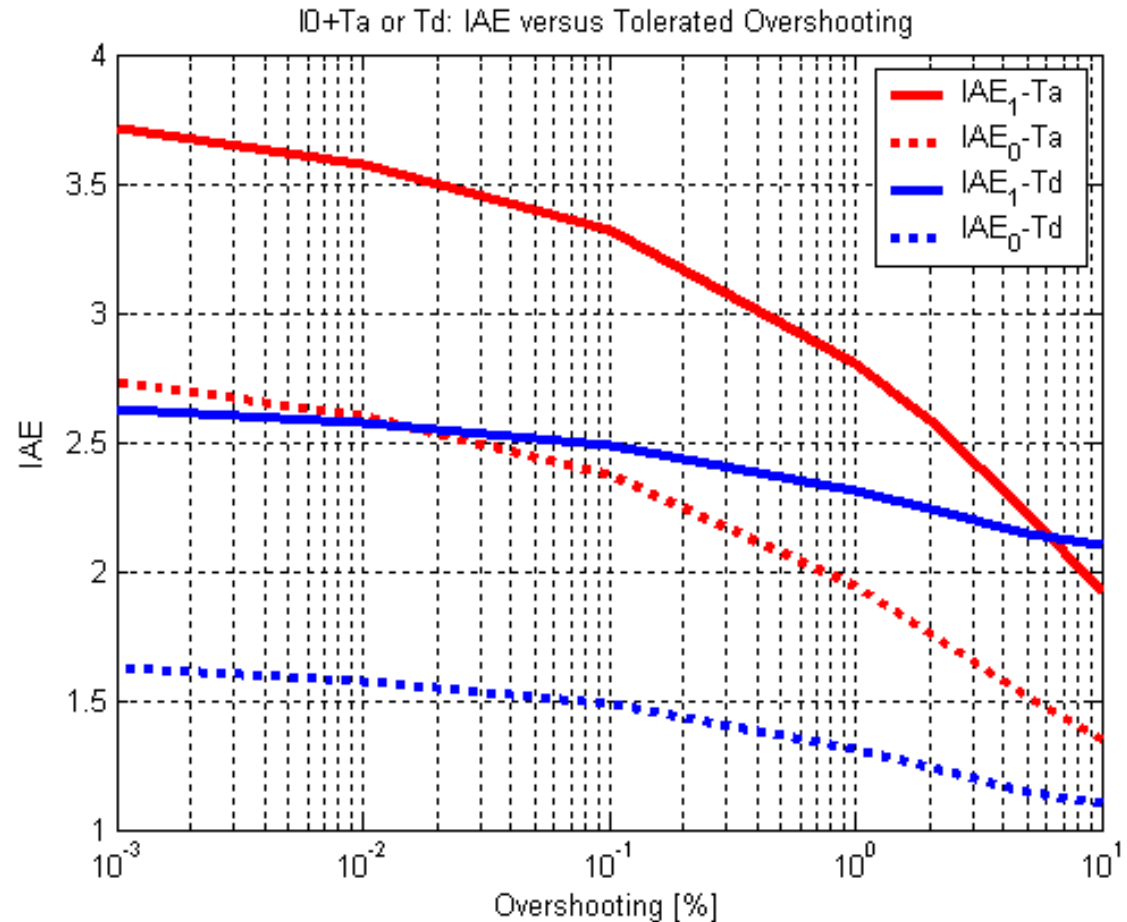


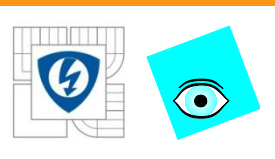
PI₀ controller: comparing loops with dead time T_d or with time constant T_a

IAE versus tolerated overshooting

For T_d it holds
 $IAE_1 = IAE_0 + 1$

For T_a this holds
just for $\varepsilon \rightarrow 0$



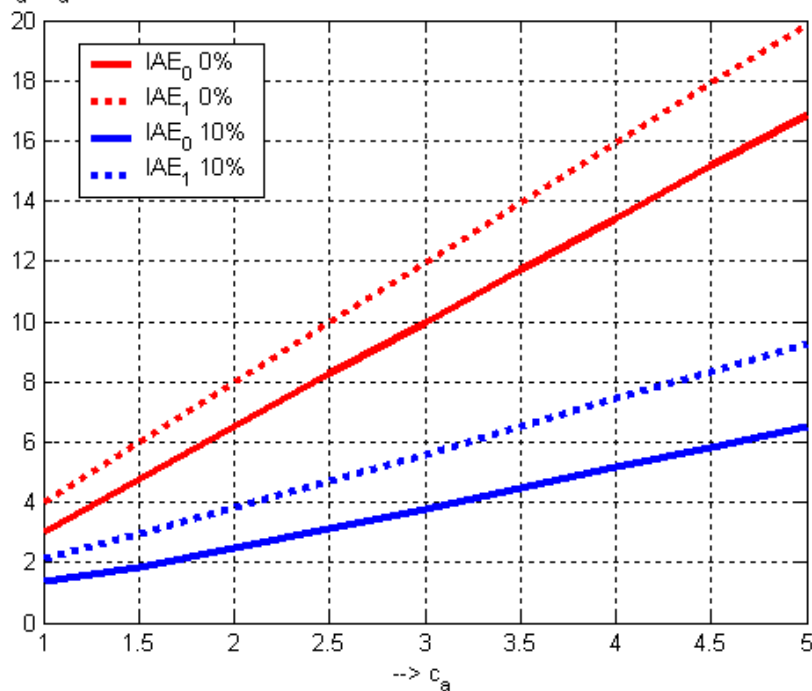


FI_0 controller - robustness characteristics

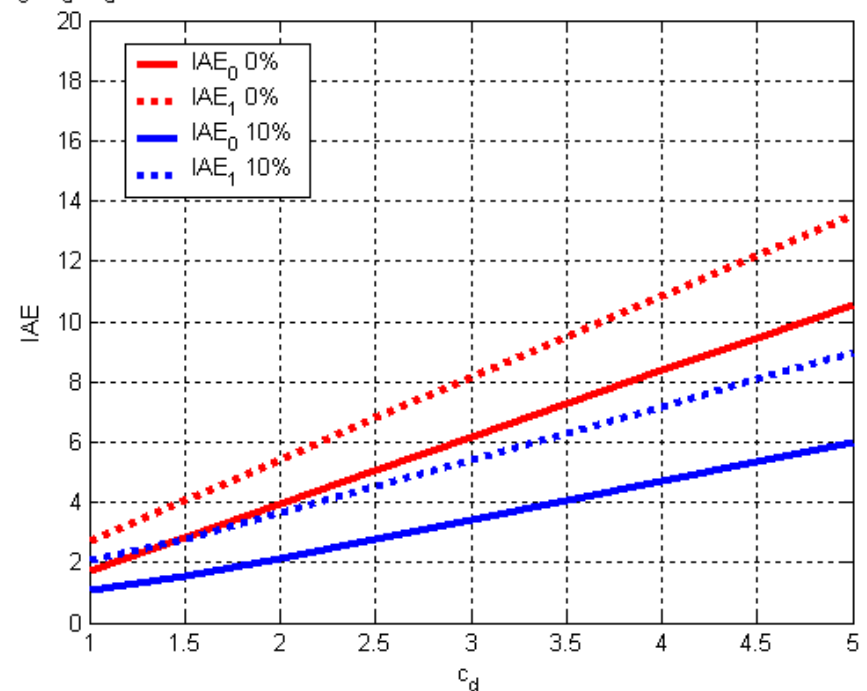
Influence of uncertainty c_a or c_d on IAE for different tolerated overshooting

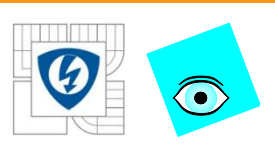
$$c_a = T_{amax}/T_{amin}, c_d = T_{dmax}/T_{dmin}$$

$I_0 + T_a, T_a$ Uncertainty Influence, Analytical MO and 10% Overshooting Tuning



$I_0 + T_d, T_d$ Uncertainty Influence, Analytical MO and 10% Overshooting Tuning



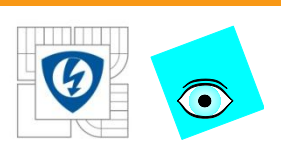


Aperiodicity border – double real dominant pole

*Oldenbourg, R.C. and H. Sartorius: Dynamik selbsttätiger Regelungen.
R.Oldenbourg-Verlag, München, 1944*

Newer references – simple formula with $\tau=2$ for overshooting 4.04%

Skogestad, S.: Simple analytic rules for model reduction and PID controller tuning.
Journal of Process Control Volume 13, Issue 4, 2003, 291-309.



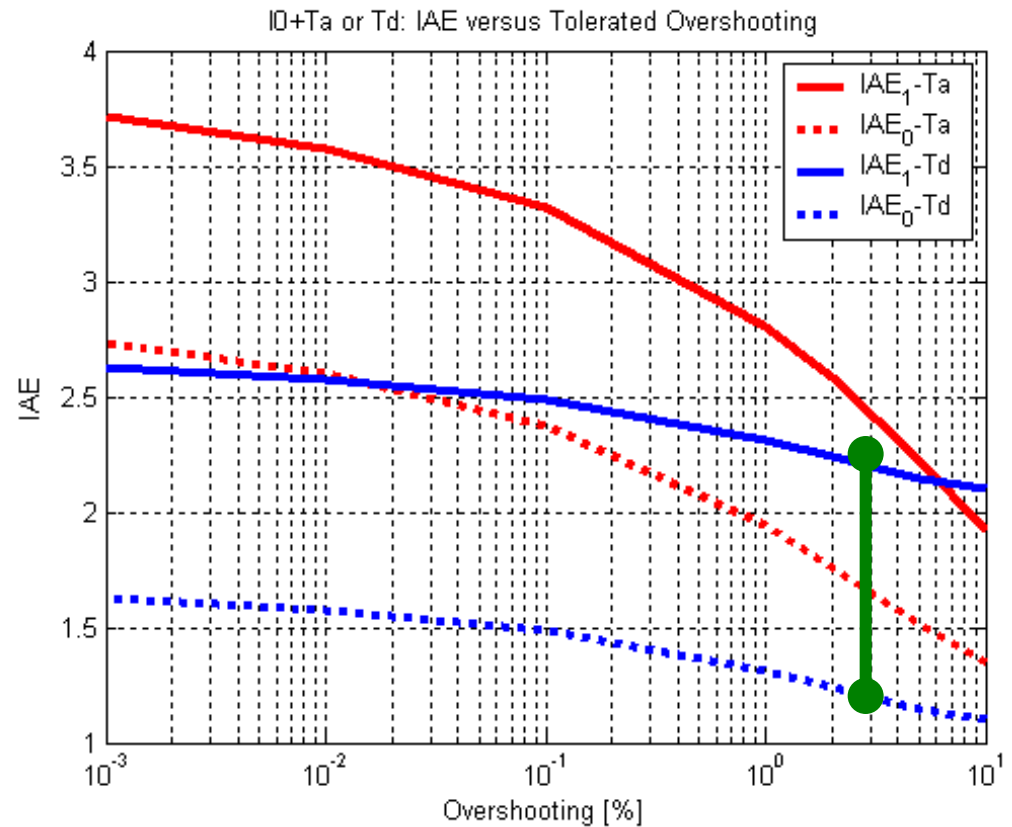
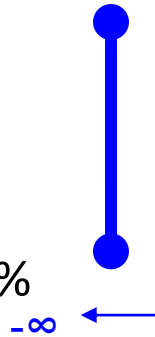
Oldenbourg a Sartorius, 1944
 T_d for overshooting 0 % given as

$$\Omega = \kappa / e$$

Skogestad, 2003

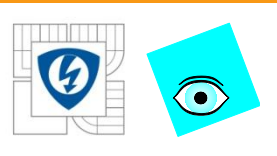
T_d for overshooting 4.04%
 given as

$$\Omega = \kappa / 2$$



$100\varepsilon_y$ [%]	10	5	4.04	2	1	0.1	0.01	0.001	0	0
$\tau = \kappa / \Omega$	1.724	1.951	2.0	2.162	2.268	2.481	2.571	2.625	2.703	2.718
$q = \Omega / \kappa$	0.580	0.515	0.5	0.465	0.441	0.403	0.389	0.381	0.37	0.368
IAE_0/T_d	1.105	1.147	1.17	1.240	1.314	1.486	1.571	1.625	1.703	1.718
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11.3.2011



I_0 controller+dead time T_d

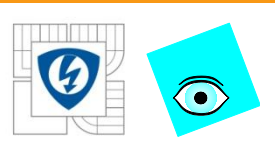
Characteristic polynomial & Critical tuning remain the same as for the FI_0 controller

Differences with respect to FI_0 control are in the:

Performance Portrait

Achievable dynamics and

Optimal tuning rules

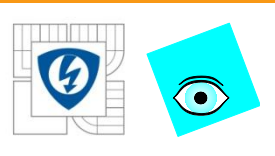


I_0 controller+dead time T_d

Closed loop transfer functions for normalized variables

$$p = T_d s ; \quad \kappa = \frac{K_0}{K} ; \quad \Omega = \frac{T_d}{T_f} ;$$

$$F_{w0}(p) = \frac{Y_0(p)}{W(p)} = (1 + p/\Omega) \frac{e^{p\Omega/\kappa}}{pe^p + \Omega/\kappa}$$
$$F_{w1}(p) = \frac{Y_1(p)}{W(p)} = (1 + p/\Omega) \frac{\Omega/\kappa}{pe^p + \Omega/\kappa}$$



I_0 controller+dead time T_d

Performance Portrait

tolerated deviations of MO

& NO ε :

0 (red)

0.00001

0.0001

0.001

0.01

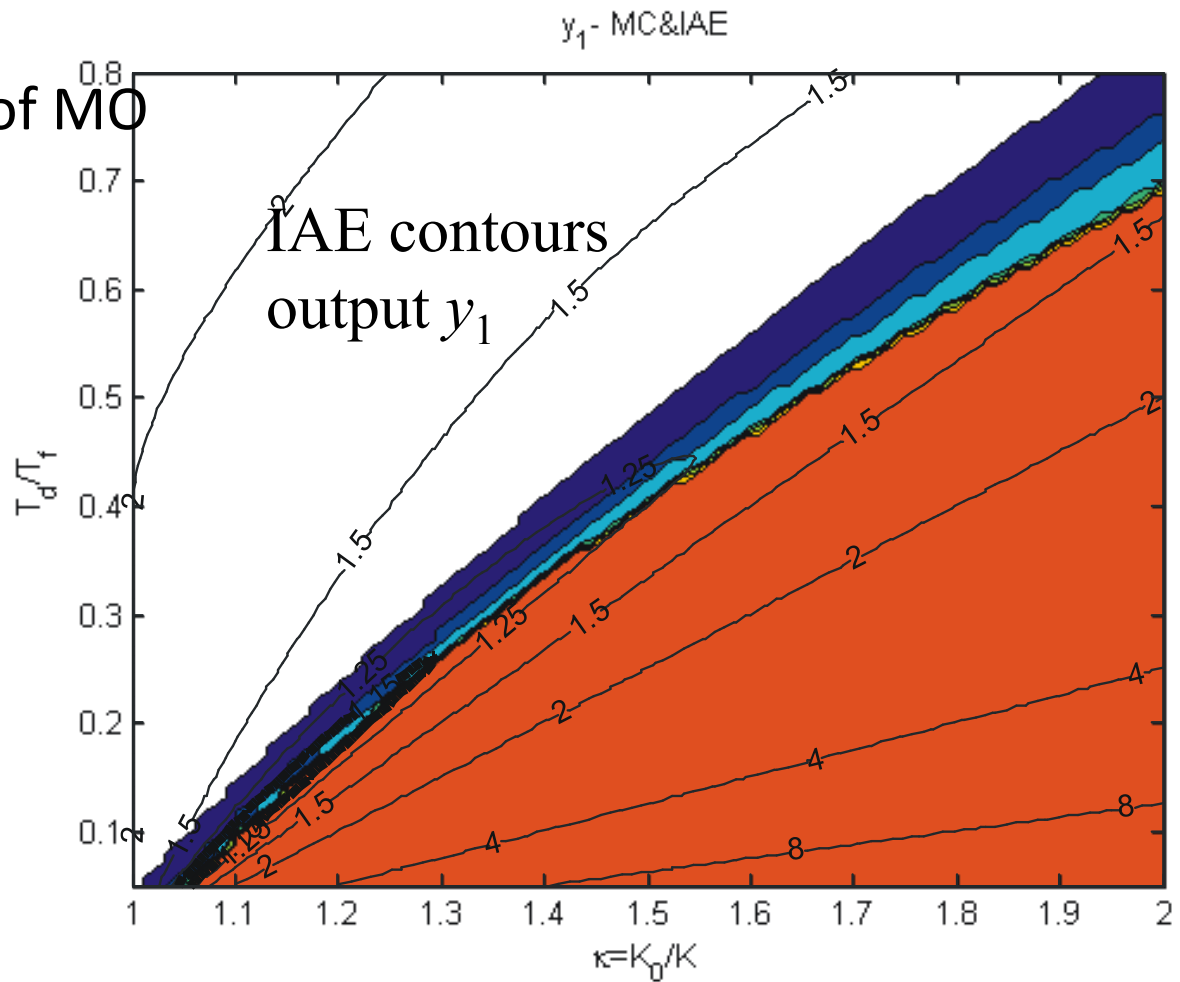
0.02

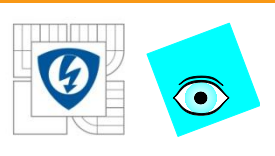
0.05

0.1 (darkblue)

with respect to

κ and T_d/T_f



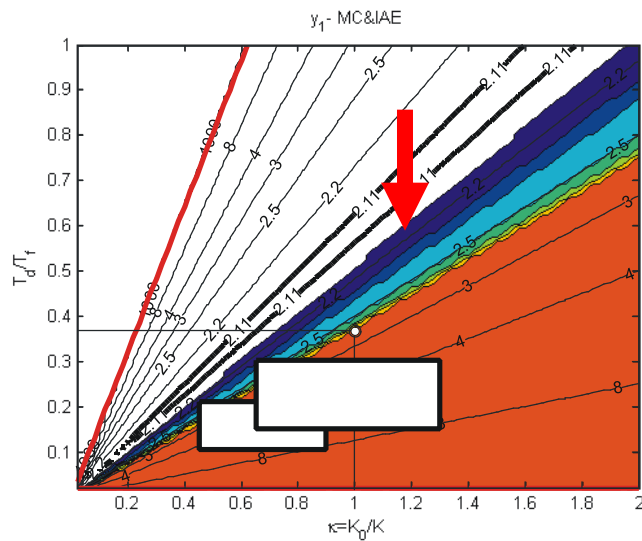


FI_0 versus I_0 controller+dead time T_d

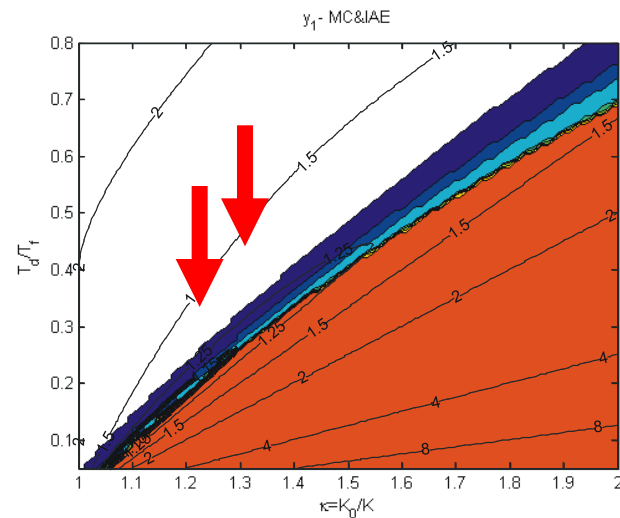
For I_0 minimal IAE values separate the area of strictly NO&MO transients from those with a tolerable overshooting = much more convenient localization of the operating point

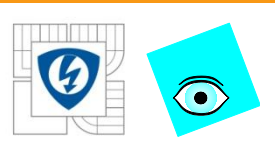
K_0 and T_f may now be tuned separately (for FI_0 in a product)

FI_0



I_0



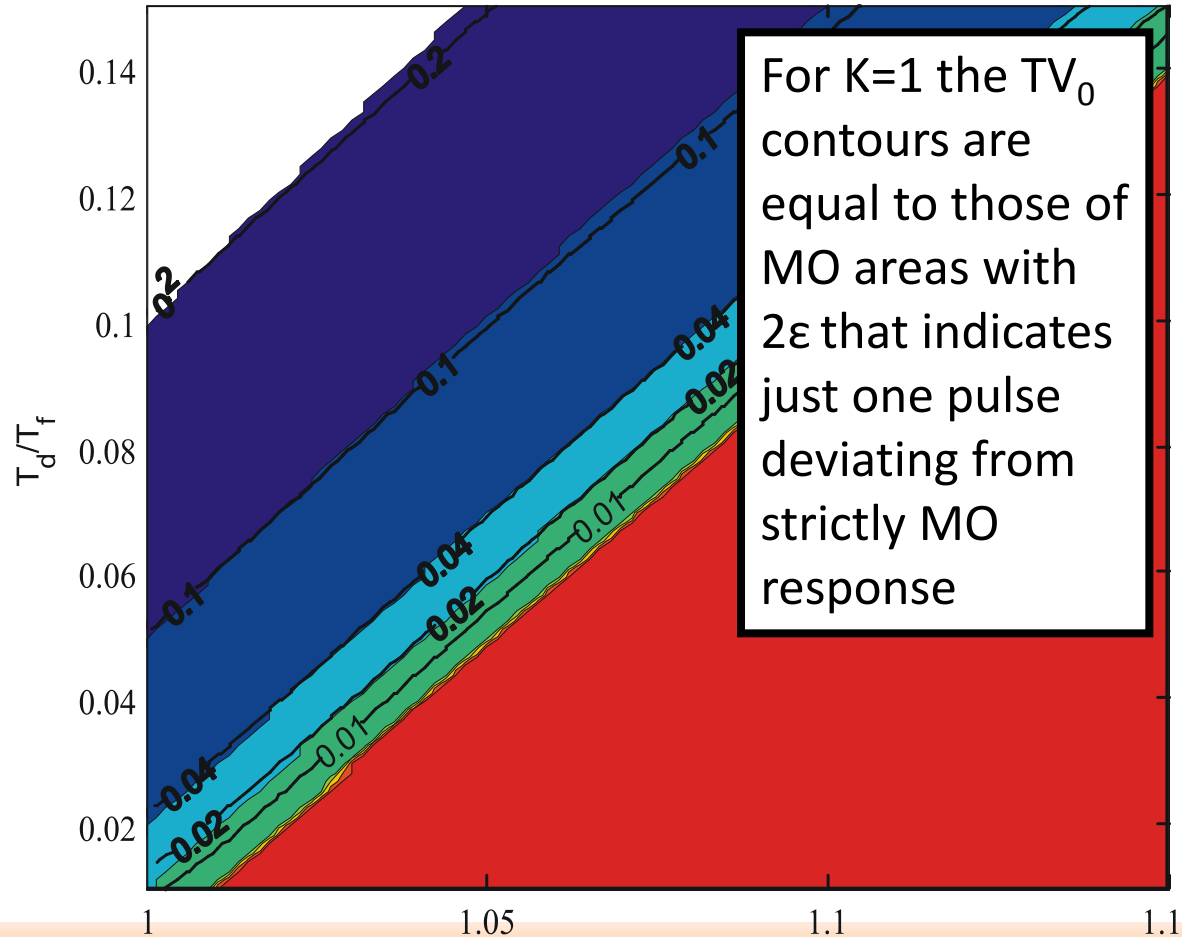


I_0 controller+dead time T_d

ε -MO vs. TV_0 contours

y_1 - MO& TV_0

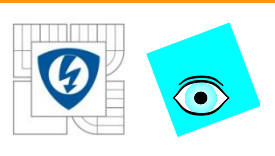
Performance
Portrait
tolerated
deviations of
MO & NO ε :
0.00001 (red)
0.0001
0.001
0.01
0.02
0.05
0.1 (darkblue)



11.3.2011

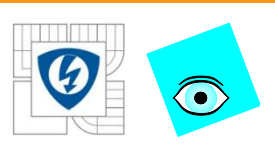
INVESTICE DO ROZVOJE VZDĚLÁVÁNÍ





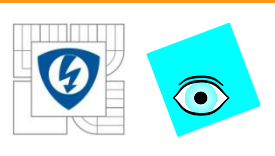
FI_0 : Conclusions

- Structure of FI_0 (I controller) = static feedforward control + DOB with the first order filter + the first order prefilter with the time constant equal to that of the DOB
- A reliable controller tuning requires approximation of the nonmodelled dynamics by time constant, or more frequently by dead time



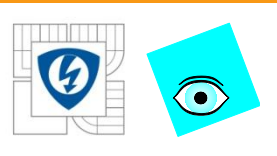
FI₀: Conclusions

- To guarantee a tolerated overshooting, robust controller tuning must respect the maximal plant gain K_{max} and the maximal value of the nonmodelled dynamics (T_{amax} , or T_{dmax})
- Sensitivity to fluctuations of the time parameter is the same as sensitivity to plant gain changes



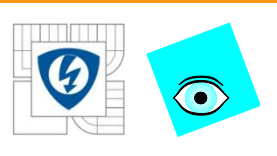
FI₀: Conclusions

- Higher requirements on control quality (lower tolerated overshooting) lead to increased sensitivity to model uncertainty
- For the time constant and lower tolerated overshooting this sensitivity is higher than for the dead time and conversely, for higher tolerated overshooting the sensitivity is higher in the case of dead time than for a time constant



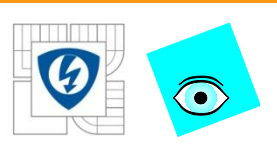
I_0 : Conclusions

- Structure of I_0 controller = static feedforward control + DOB with the first order filter
- In comparing with FI_0 , the setpoint responses may be reasonably improved
- A reliable controller tuning requires approximation of the nonmodelled dynamics (by a time constant, or more frequently by a dead time).



I₀: Conclusions

- To guarantee a tolerated overshooting, robust controller tuning must respect the maximal plant gain K_{max} and the maximal value of the nonmodeled dynamics (T_{amax} , or T_{dmax})
- Sensitivity to fluctuations of the time parameter is different from the sensitivity to plant gain changes.



I_0 : Conclusions

- Requirement to keep the disturbance response (i.e. $T_f K_0 = \text{const}$) leads finally to dynamics comparable with those achieved by reasonably simpler FI_0
- Therefore, we will deal mostly just with the simpler $FI_0 = I$ -controller